

## QUAD DIFFERENTIAL PECL RECEIVERS

### FEATURES

- Low-Voltage Functional Replacements for the Agere BRF1A, BRF2A, BRS2A, and BRS2B
- Pin-Equivalent to General Trade 26LS32 Devices
- High-Input Impedance Approximately 8 k $\Omega$
- 3.5-ns Maximum Propagation Delay
- TB3R1 Provides 50-mV Hysteresis
- TB3R2 With -125-mV Threshold Offset for Preferred State Output
- -0.5-V to 5.2-V Common Mode Range
- Single 3.3 V  $\pm 10\%$  Supply
- Slew Rate Limited (0.5 ns min 80% to 20%)
- TB3R2 Output Defaults to Logic 1 When Inputs Left Open or Shorted to V<sub>CC</sub> or GND
- ESD Protection HBM > 3 kV, CDM > 2 kV
- Operating Temperature Range: -40°C to 85°C
- Available SOIC (D) Package

### APPLICATIONS

- Digital Data or Clock Transmission Over Balanced Lines

### DESCRIPTION

These quad differential receivers accept digital data over balanced transmission lines. They translate differential input logic levels to TTL output logic levels.

The TB3R1 is a pin- and function-compatible replacement for the Agere Systems BRF1A and BRF2A; it includes 3-kV HBM and 2-kV CDM ESD protection.

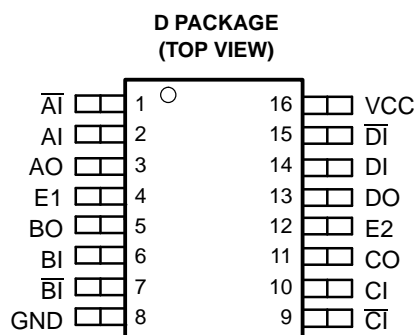
The TB3R2 is a pin- and function-compatible replacement for the Agere Systems BRS2A and BRS2B and incorporates a -125-mV receiver input offset, preferred state output, 3-kV HBM and 2-kV CDM ESD protection. The TB3R2 preferred state feature places the output in the high state when the inputs are open, shorted to ground, or shorted to the power supply.

The power-down loading characteristics of the receiver input circuit are approximately 8 k $\Omega$  relative to the power supplies; hence they do not load the transmission line when the circuit is powered down.

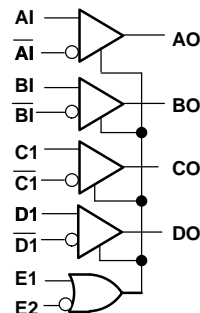
The package for these differential line receivers is the 16-pin SOIC (D) package.

The enable inputs of this device include internal pullup resistors of approximately 40 k $\Omega$  that are connected to V<sub>CC</sub> to ensure a logical high level input if the inputs are open circuited.

### PIN ASSIGNMENTS



### FUNCTIONAL BLOCK DIAGRAM



### ENABLE TRUTH TABLE

E1	E2	CONDITION
0	0	Active
1	0	Active
0	1	Disabled
1	1	Active



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### ORDERING INFORMATION

PART NUMBER	PART MARKING	Package	LEAD FINISH	STATUS
TB3R1D	TB3R1	SOIC	NiPdAu	Production
TB3R2D	TB3R2	SOIC	NiPdAu	Production
TB3R1LD	TB3R1	SOIC	SnPb	Production
TB3R2LD	TB3R2	SOIC	SnPb	Production

### POWER DISSIPATION RATINGS

PACKAGE	CIRCUIT BOARD MODEL	POWER RATING $T_A \leq 25^\circ\text{C}$	THERMAL RESISTANCE, JUNCTION-TO-AMBIENT WITH NO AIR FLOW	DERATING FAC- TOR <sup>(1)</sup> $T_A \geq 25^\circ\text{C}$	POWER RATING $T_A = 85^\circ\text{C}$
D	Low-K <sup>(1)</sup>	763 mW	131.1 $^\circ\text{C/W}$	7.6 mW/ $^\circ\text{C}$	305 mW
	High-K <sup>(2)</sup>	1190 mW	84.1 $^\circ\text{C/W}$	11.9 mW/ $^\circ\text{C}$	475 mW
DW	Low-K <sup>(1)</sup>	831 mW	120.3 $^\circ\text{C/W}$	8.3 mW/ $^\circ\text{C}$	332 mW
	High-K <sup>(2)</sup>	1240 mW	80.8 $^\circ\text{C/W}$	12.4 mW/ $^\circ\text{C}$	494 mW

(1) In accordance with the low-K thermal metric definitions of EIA/JESD51-3.

(2) In accordance with the high-K thermal metric definitions of EIA/JESD51-7.

### THERMAL CHARACTERISTICS

PARAMETER		PACKAGE	VALUE	UNIT
$\theta_{JB}$	Junction-to-Board Thermal Resistance	D	47.5	$^\circ\text{C/W}$
		DW	53.7	$^\circ\text{C/W}$
$\theta_{JC}$	Junction-to-Case Thermal Resistance	D	44.2	$^\circ\text{C/W}$
		DW	47.1	$^\circ\text{C/W}$

### ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted<sup>(1)</sup>

			UNIT
Supply voltage, $V_{CC}$			0 V to 6 V
Magnitude of differential bus (input) voltage, $ V_{AI} - V_I $ , $ V_{BI} - V_I $ , $ V_{CI} - V_I $ , $ V_{DI} - V_I $			6.5 V
ESD	Human Body Model <sup>(2)</sup>	All pins	$\pm 3$ kV
	Charged-Device Model <sup>(3)</sup>	All pins	$\pm 2$ kV
Continuous power dissipation			See Dissipation Rating Table
Storage temperature, $T_{stg}$			$-65^\circ\text{C}$ to $150^\circ\text{C}$

(1) Stresses beyond those listed under „absolute maximum ratings“ may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under „recommended operating conditions“ is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) Tested in accordance with JEDEC Standard 22, Test Method A114-A.

(3) Tested in accordance with JEDEC Standard 22, Test Method C101.

## RECOMMENDED OPERATING CONDITIONS

	MIN	Nom	MAX	UNIT
Supply voltage, $V_{CC}$	3	3.3	3.6	V
Bus pin input voltage, $V_{AI}$ , $V$ , $V_{BI}$ , $V$ , $V_{CI}$ , $V$ , $V_{DI}$ , $V$	-0.6 <sup>(1)</sup>		5.3	V
Magnitude of differential input voltage, $ V_{AI} - V $ , $ V_{BI} - V $ , $ V_{CI} - V $ , $ V_{DI} - V $	0.1		5	V
Operating free-air temperature, $T_A$	-40		85	°C

- (1) The algebraic convention, in which the least positive (most negative) limit is designated as minimum is used in this data sheet, unless otherwise noted.

## DEVICE ELECTRICAL CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{CC}$ Supply current <sup>(1)</sup>	Outputs disabled			34	mA
	Outputs enabled			32	mA

- (1) Current is dc power draw as measured through GND pin and does not include power delivered to load.

## ELECTRICAL CHARACTERISTICS

over operating free-air temperature range unless otherwise noted

parameter	test conditions	min	typ	max	unit
$V_{OL}$ Output low voltage	$V_{CC} = 3\text{ V}$ , $I_{OL} = 8\text{ mA}$			0.4	V
$V_{OH}$ Output high voltage	$V_{CC} = 3\text{ V}$ , $I_{OH} = -400\text{ }\mu\text{A}$	2.4			V
$V_{IL}$ Low level enable input voltage <sup>(1)</sup>	$V_{CC} = 3.6\text{ V}$			0.8	V
$V_{IH}$ High level enable input voltage <sup>(1)</sup>	$V_{CC} = 3.6\text{ V}$	2			V
$V_{IK}$ Enable input clamp voltage	$V_{CC} = 3\text{ V}$ , $I_I = -5\text{ mA}$			-1 <sup>(2)</sup>	V
$V_{TH+}$ Positive-going differential input threshold voltage <sup>(1)</sup> , ( $V_{XI} - V$ )	$x = A, B, C, \text{ or } D$	TB3R1		100	mV
		TB3R2 <sup>(3)</sup>		-50	mV
$V_{TH-}$ Negative-going differential input threshold voltage <sup>(1)</sup> , ( $V_{XI} - V$ )	$x = A, B, C, \text{ or } D$	TB3R1		-100 <sup>(2)</sup>	mV
		TB3R2 <sup>(3)</sup>		-200 <sup>(2)</sup>	mV
$V_{HYST}$ Differential input threshold voltage hysteresis, ( $V_{TH+} - V_{TH-}$ )	TB3R1		50		mV
$I_{OZL}$ Output off-state current, (High-Z)	$V_{CC} = 3.6\text{ V}$	$V_O = 0.4\text{ V}$		-20 <sup>(2)</sup>	$\mu\text{A}$
		$V_O = 2.4\text{ V}$		20	$\mu\text{A}$
$I_{OS}$ Output short circuit current <sup>(4)</sup>	$V_{CC} = 3.6\text{ V}$			-100 <sup>(2)</sup>	mA
$I_{IL}$ Enable input low current	$V_{CC} = 3.6\text{ V}$ , $V_{IN} = 0.4\text{ V}$			-400 <sup>(2)</sup>	$\mu\text{A}$
$I_{IH}$ Enable input high current	$V_{CC} = 3.6\text{ V}$	$V_{IN} = 2.7\text{ V}$		20	$\mu\text{A}$
		$V_{IN} = 3.6\text{ V}$		100	$\mu\text{A}$
$I_{IL}$ Differential input low current	$V_{CC} = 3.6\text{ V}$ , $V_{IN} = -1.2\text{ V}$			-2 <sup>(2)</sup>	mA
$I_{IH}$ Differential input high current	$V_{CC} = 3.6\text{ V}$ , $V_{IN} = 5.3\text{ V}$			1	mA
$R_O$ Output resistance			20		$\Omega$

- (1) The input levels and difference voltage provide no noise immunity and should be tested only in a static, noise-free environment.  
(2) This parameter is listed using a magnitude and polarity/direction convention, rather than an algebraic convention, to match the original Agere data sheet.  
(3) Outputs of unused receivers assume a logic 1 level when the inputs are left open. (It is recommended that all unused positive inputs be tied to the positive power supply. No external series resistor is required.)  
(4) Test must be performed one lead at a time to prevent damage to the device.

**SWITCHING CHARACTERISTICS**

over operating free-air temperature range unless otherwise noted

parameter		test conditions	min	typ	max	unit
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 0 \text{ pF}^{(1)}$ , See Figure 2 and Figure 4	1.8	3.5		ns
$t_{PHL}$	Propagation delay time, high-to-low-level output		1.8	3.5		
$t_{PLH}$	Propagation delay time, low-to-high-level output	$C_L = 15 \text{ pF}$ , See Figure 2 and Figure 4	2.3	4		ns
$t_{PHL}$	Propagation delay time, high-to-low-level output		2.3	4		
$t_{PHZ}$	Output disable time, high-level-to-high-impedance output <sup>(2)</sup>	$C_L = 5 \text{ pF}$ See Figure 3 and Figure 5	4.4	12		ns
$t_{PLZ}$	Output disable time, low-level-to-high-impedance output <sup>(2)</sup>		3.3	12		ns
$t_{skew1}$	Pulse width distortion, $ t_{PHL} - t_{PLH} $	$C_L = 10 \text{ pF}$ , See Figure 2 and Figure 4		0.7		ns
		$C_L = 150 \text{ pF}$ , See Figure 2 and Figure 4		4		ns
$\Delta t_{skew1p-p}$	Part-to-part output waveform skew <sup>(3)</sup>	$C_L = 10 \text{ pF}$ , $T_A = 75^\circ\text{C}$ , See Figure 2 and Figure 4	0.8	1.4		ns
		$C_L = 10 \text{ pF}$ , $T_A = -40^\circ\text{C}$ to $85^\circ\text{C}$ , See Figure 2 and Figure 4		1.5		ns
$\Delta t_{skew}$	Same part output waveform skew <sup>(3)</sup>	$C_L = 10 \text{ pF}$ , See Figure 2 and Figure 4		0.3		ns
$t_{PZH}$	Output enable time, high-impedance-to-high-level output <sup>(4)</sup>	$C_L = 10 \text{ pF}$ , See Figure 3 and Figure 4	6	12		ns
$t_{PZL}$	Output enable time, high-impedance-to-low-level output <sup>(4)</sup>		4	12		ns
$t_{TLH}$	Rise time (20%-80%)	$C_L = 10 \text{ pF}$ , See Figure 2 and Figure 4	0.5		2	ns
$t_{THL}$	Fall time (80%-20%)		0.5		2	ns

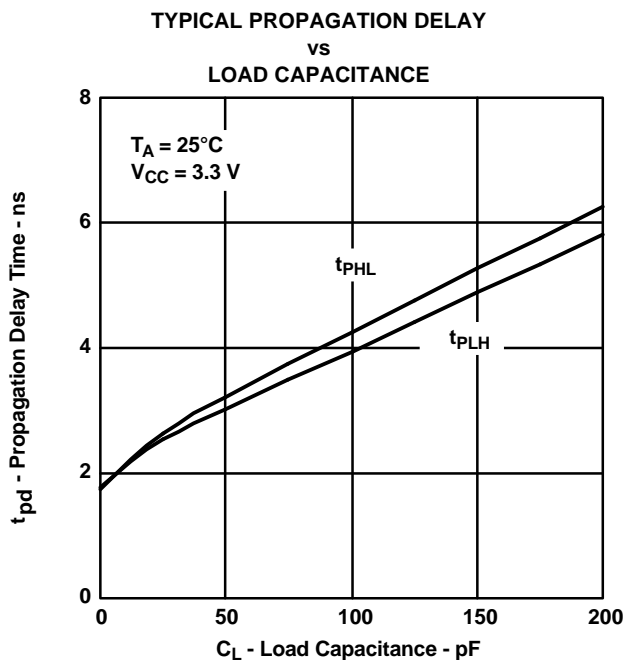
(1) The propagation delay values with a 0 pF load are based on design and simulation.

(2) See Table 1.

(3) Output waveform skews are when devices operate with the same supply voltage, same temperature, have the same packages and the same test circuits.

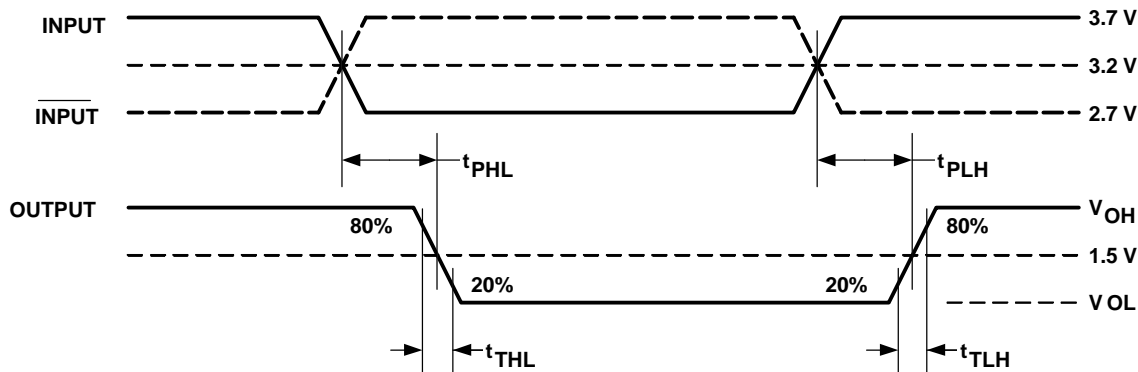
(4) See Table 1.

## TYPICAL CHARACTERISTICS

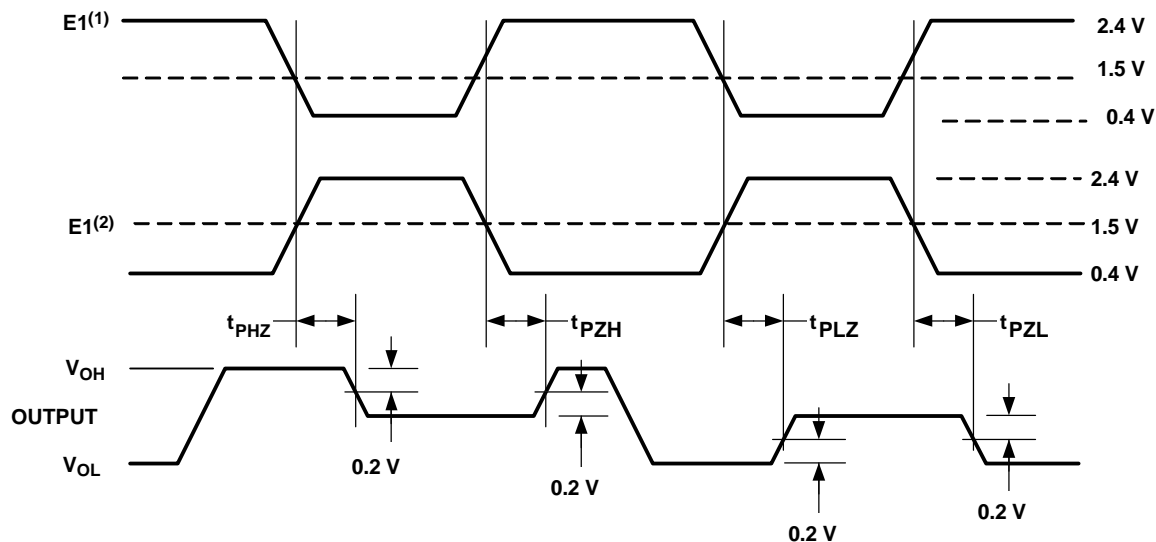


- A. NOTE: This graph is included as an aid to the system designers. Total circuit delay varies with load capacitance. The total delay is the sum of the delay due to external capacitance and the intrinsic delay of the device. Intrinsic delay is listed in the table above as the 0 pF load condition. The incremental increase in delay between the 0 pF load condition and the actual total load capacitance represents the extrinsic, or external delay contributed by the load.

**Figure 1. Typical Propagation Delay vs Load Capacitance at 25°C**



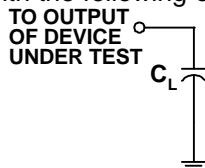
**Figure 2. Receiver Propagation Delay Times**

**TYPICAL CHARACTERISTICS (continued)**

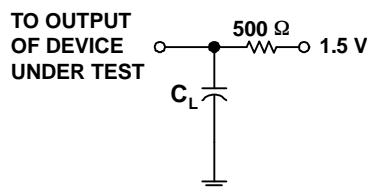
- A. E2 = 1 while E1 changes states.
- B. E1 = 0 while E2 changes states.

**Figure 3. Receiver Enable and Disable Timing**

Parametric values specified under the Electrical Characteristics and Timing Characteristics sections for the data transmission driver devices are measured with the following output load circuits.



$C_L$  includes test-fixture and probe capacitance.

**Figure 4. Receiver Propagation Delay Time and Enable Time (t<sub>PZH</sub>, t<sub>PZL</sub>) Test Circuit**

$C_L$  includes test-fixture and probe capacitance.

**Figure 5. Receiver Disable Time (t<sub>PHZ</sub>, t<sub>PLZ</sub>) Test Circuit**

# TYPICAL CHARACTERISTICS (continued)

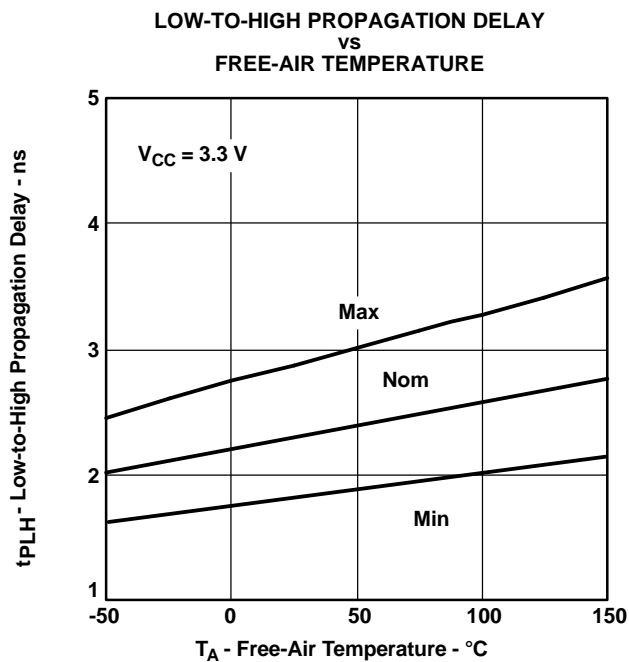


Figure 6.

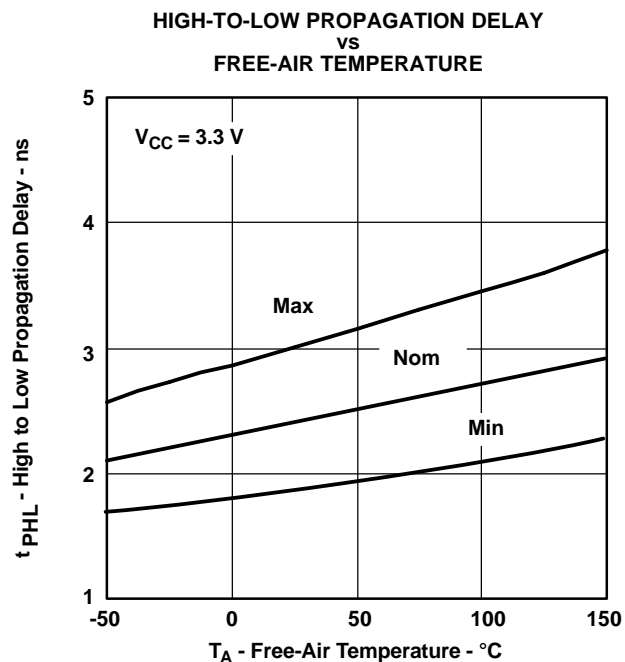


Figure 7.

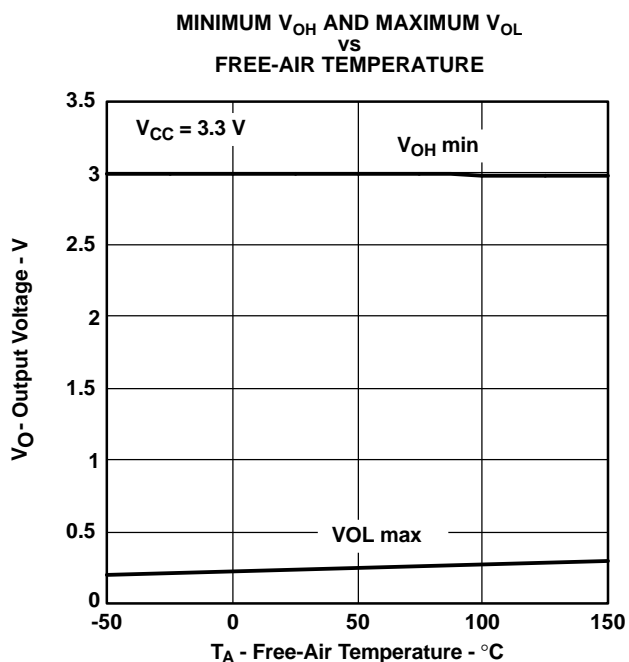


Figure 8.

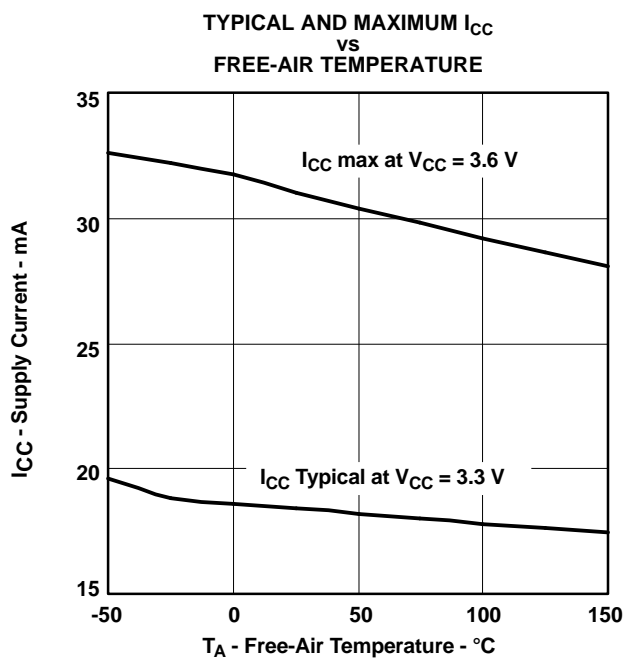


Figure 9.

## APPLICATION INFORMATION

## Power Dissipation

The power dissipation rating, often listed as the package dissipation rating, is a function of the ambient temperature,  $T_A$ , and the airflow around the device. This rating correlates with the device's maximum junction temperature, sometimes listed in the absolute maximum ratings tables. The maximum junction temperature accounts for the processes and materials used to fabricate and package the device, in addition to the desired life expectancy.

There are two common approaches to estimating the internal die junction temperature,  $T_J$ . In both of these methods, the device internal power dissipation  $P_D$  needs to be calculated. This is done by totaling the supply power(s) to arrive at the system power dissipation:

$$\sum V_{Sn} \times I_{Sn} \quad (1)$$

and then subtracting the total power dissipation of the external load(s):

$$\sum (V_{Ln} \times I_{Ln}) \quad (2)$$

The first  $T_J$  calculation uses the power dissipation and ambient temperature, along with one parameter:  $\theta_{JA}$ , the junction-to-ambient thermal resistance, in degrees Celsius per watt.

The product of  $P_D$  and  $\theta_{JA}$  is the junction temperature rise above the ambient temperature. Therefore:

$$T_J = T_A \sum [P_D \times \theta_{JA}] \quad (3)$$

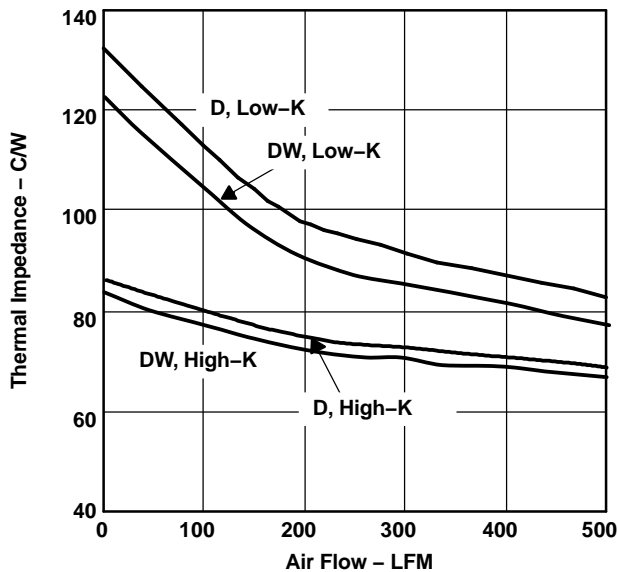


Figure 10. Thermal Impedance vs Air Flow

Note that  $\theta_{JA}$  is highly dependent on the PCB on which the device is mounted, and on the airflow over the device and PCB. JEDEC/EIA has defined standardized test conditions for measuring  $\theta_{JA}$ . Two commonly used conditions are the low-K and the high-K boards, covered by EIA/JESD51-3 and EIA/JESD51-7 respectively. Figure 10 shows the low-K and high-K values of  $\theta_{JA}$  versus air flow for this device and its package options.

The standardized  $\theta_{JA}$  values may not accurately represent the conditions under which the device is used. This can be due to adjacent devices acting as heat sources or heat sinks, to nonuniform airflow, or to the system PCB having significantly different thermal characteristics than the standardized test PCBs. The second method of system thermal analysis is more accurate. This calculation uses the power dissipation and ambient temperature, along with two device and two system-level parameters:

- $\theta_{JC}$ , the junction-to-case thermal resistance, in degrees Celsius per watt
- $\theta_{JB}$ , the junction-to-board thermal resistance, in degrees Celsius per watt
- $\theta_{CA}$ , the case-to-ambient thermal resistance, in degrees Celsius per watt
- $\theta_{BA}$ , the board-to-ambient thermal resistance, in degrees Celsius per watt.

In this analysis, there are two parallel paths, one through the case (package) to the ambient, and another through the device to the PCB to the ambient. The system-level junction-to-ambient thermal impedance,  $\theta_{JA(S)}$ , is the equivalent parallel impedance of the two parallel paths:

$$T_J = T_A \sum [P_D \times \theta_{JA(S)}] \quad (4)$$

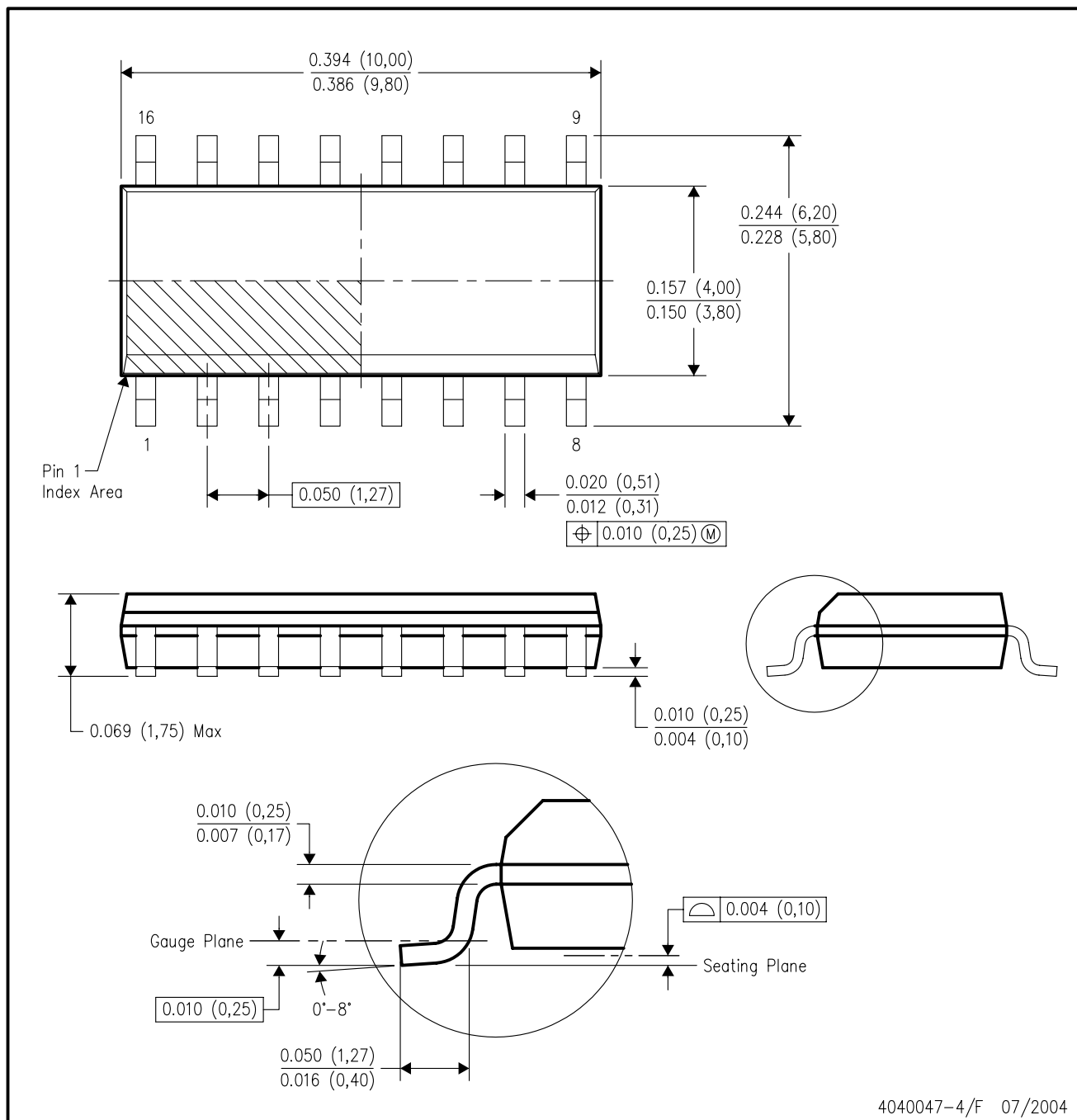
where

$$\theta_{JA(S)} = \frac{[\theta_{JC} \sum \theta_{CA} \times \theta_{JB} \sum \theta_{BA}]}{\theta_{JC} \sum \theta_{CA} \sum \theta_{JB} \sum \theta_{BA}} \quad (5)$$

The device parameters  $\theta_{JC}$  and  $\theta_{JB}$  account for the internal structure of the device. The system-level parameters  $\theta_{CA}$  and  $\theta_{BA}$  take into account details of the PCB construction, adjacent electrical and mechanical components, and the environmental conditions including airflow. Finite element (FE), finite difference (FD), or computational fluid dynamics (CFD) programs can determine  $\theta_{CA}$  and  $\theta_{BA}$ . Details on using these programs are beyond the scope of this data sheet, but are available from the software manufacturers.

## D (R-PDSO-G16)

## PLASTIC SMALL-OUTLINE PACKAGE



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