

TLV320x 40ns, microPOWER, Push-Pull Output Comparators

1 Features

- Low Propagation Delay: 40ns
- Low Quiescent Current: 40µA per Channel
- Input Common-Mode Range Extends 200mV Beyond Either Rail
- Low Input Offset Voltage: 1mV
- Push-Pull Outputs
- Supply Range: 2.7V to 5.5V
- Industrial Temperature Range: –40°C to 125°C
- Small Packages: 5-Pin SC70, 5-Pin SOT-23, 8-Pin SOIC, 8-Pin VSSOP

2 Applications

- Inspection Equipment
- Test and Measurement
- High-Speed Sampling Systems
- Telecom
- Portable Communications

3 Description

The TLV3201 and TLV3202 are single- and dual-channel comparators that offer the ultimate combination of high speed (40ns) and low-power consumption (40µA), all in extremely small packages with features such as rail-to-rail inputs, low offset voltage (1mV), and large output drive current. The devices are also very easy to implement in a wide variety of applications where response time is critical.

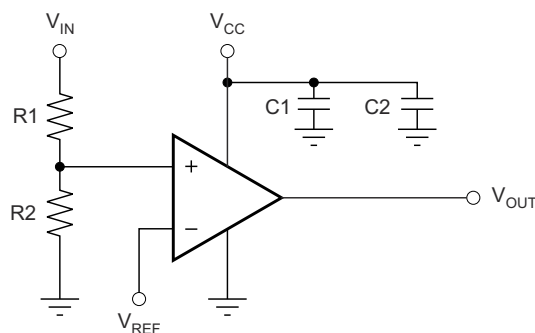
The TLV320x family is available in single (TLV3201) and dual (TLV3202) channel versions, both with push-pull outputs. The TLV3201 is available in 5-pin SOT-23 and 5-pin SC70 packages. The TLV3202 is available in 8-pin SOIC and 8-pin VSSOP packages. All devices are specified for operation across the expanded industrial temperature range of –40°C to 125°C.

Device Information

PART NUMBER	PACKAGE ⁽¹⁾	BODY SIZE (NOM) ⁽²⁾
TLV3201	SOT-23 (5)	2.90mm × 1.60mm
	SC70 (5)	2.00mm × 1.25mm
TLV3202	VSSOP (8)	3.00mm × 3.00mm
	SOIC (8)	4.90mm × 3.91mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

(2) The package size (length × width) is a nominal value and includes pins, where applicable



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Threshold Detector



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4 Device Comparison Table

DEVICE	DESCRIPTION
TLV3011	5µA (maximum) open-drain, 1.8V to 5.5V with integrated voltage reference in 1.5mm × 1.5mm micro-sized packages
TLV3012	5µA (maximum) push-pull, 1.8V to 5.5V with integrated voltage reference in micro-sized packages
TLV3501	4.5ns, rail-to-rail, push-pull comparator in micro-sized packages
LMV7235	75ns, 65µA, 2.7V to 5.5V, rail-to-rail input comparator with open-drain output
REF3333	30ppm/°C drift, 3.9µA, SOT23-3, SC70-3 voltage reference

5 Pin Configuration and Functions

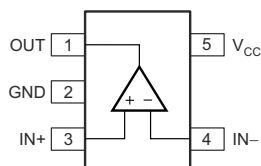


Figure 5-1. TLV3201 DCK and DBV Packages 5-Pin SC70-5 and SOT-23 Top View

Pin Functions: TLV3201

PIN		I/O	DESCRIPTION
NAME	NO.		
GND	2	—	Negative supply, ground
IN–	4	I	Negative input
IN+	3	I	Positive input
OUT	1	O	Output
V _{CC}	5	—	Positive supply

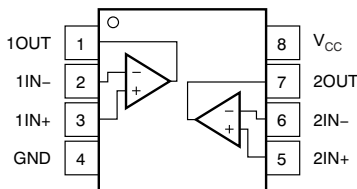


Figure 5-2. TLV3202 D and DGK Packages 8-Pin SOIC and VSSOP Top View

Pin Functions: TLV3202

PIN		I/O	DESCRIPTION
NAME	NO.		
1IN–	2	I	Negative input, comparator 1
1IN+	3	I	Positive input, comparator 1
1OUT	1	O	Output, comparator 1
2IN–	6	I	Negative input, comparator 2
2IN+	5	I	Positive input, comparator 2
2OUT	7	O	Output, comparator 2
GND	4	—	Negative supply, ground
V _{CC}	8	—	Positive supply

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Voltage	Supply voltage		7	V
	Signal input pins ⁽²⁾	–0.5	(V _{CC}) + 0.5	
Current	Signal input pins ⁽²⁾	–10	10	mA
	Output short circuit ⁽³⁾		100	
Temperature	Operating	–55	125	°C
	Junction, T _J		150	
	Storage, T _{stg}	–65	150	

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under [Section 6.3](#). Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Input pins are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails must be current limited to 10 mA or less.
- (3) Short-circuit to ground.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD) Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾		±2000	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	TLV3201	±2000	
		TLV3202	±1000	

- (1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V _S	Supply voltage, V _S = (V _{S+}) – (V _{S–})	2.7 (±1.35)	5.5 (±2.75)	V
	Specified temperature	–40	125	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		TLV3201		TLV3202		UNIT
		DBV (SOT-23)	DCK (SC70)	D (SOIC)	DGK (VSSOP)	
		5 PINS	5 PINS	8 PINS	8 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	237.8	281.9	143.6	201.9	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	108.7	97.6	97.2	92.5	°C/W
R _{θJB}	Junction-to-board thermal resistance	64.1	68.3	84.2	123.3	°C/W
ψ _{JT}	Junction-to-top characterization parameter	12.1	2.6	45.5	23.0	°C/W
ψ _{JB}	Junction-to-board characterization parameter	63.3	67.3	83.7	212.6	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Electrical Characteristics: $V_{CC} = 5V$

at $T_A = 25^\circ C$ and $V_{CC} = 5V$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
OFFSET VOLTAGE							
V _{IO-TLV3201}	Input offset voltage TLV3201	V _{CM} = V _{CC} / 2		1	3	mV	
		T _A = −40°C to 125°C			4		
V _{IO-TLV3202}	Input offset voltage TLV3202	V _{CM} = V _{CC} / 2		1	5	mV	
		T _A = −40°C to 125°C			6		
dV _{OS} /dT	Input offset voltage drift	T _A = −40°C to 125°C		1	10	μV/°C	
PSRR	Power-supply rejection ratio	V _{CM} = V _{CC} / 2, V _{CC} = 2.5V to 5.5V	65	85		dB	
	Input hysteresis			1.2		mV	
INPUT BIAS CURRENT							
I _{IB}	Input bias current	V _{CM} = V _{CC} / 2		1	50	pA	
		T _A = −40°C to 125°C			5	nA	
I _{IO}	Input offset current	V _{CM} = V _{CC} / 2		1	50	pA	
		T _A = −40°C to 125°C			2.5	nA	
INPUT VOLTAGE RANGE							
V _{CM}	Common-mode voltage	T _A = −40°C to 125°C	(V _{EE}) − 0.2		(V _{CC}) + 0.2	V	
CMRR	Common-mode rejection ratio	−0.2V < V _{CM} < 5.2V	60	70		dB	
INPUT IMPEDANCE							
	Common mode			10 ¹³ 2		Ω pF	
	Differential			10 ¹³ 4		Ω pF	
OUTPUT							
V _{OL}	Voltage output swing from lower rail	I _{SINK} = 4mA		175	190	mV	
		T _A = −40°C to 125°C			225		
V _{OH}	Voltage output swing from upper rail	I _{SOURCE} = 4mA		120	140	mV	
		T _A = −40°C to 125°C			170		
I _{SC}	Short-circuit current (per comparator)	I _{SC} sinking	40	48		mA	
		T _A = −40°C to 125°C	See Figure 6-14				
		I _{SC} sourcing	52	60			
		T _A = −40°C to 125°C	See Figure 6-14				
POWER SUPPLY							
V _{CC}	Specified voltage		2.7		5.5	V	
I _Q	Quiescent current	T _A = 25°C		40	50	μA	
		T _A = −40°C to 125°C			65		

6.6 Switching Characteristics: $V_{CC} = 5V$

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
t _{PD}	Propagation delay time	Low to high	Input overdrive = 20mV , C _L = 15pF		47	50	ns
			Input overdrive = 100mV , C _L = 15pF		43	50	
			T _A = −40°C to 125°C			55	
		High to low	Input overdrive = 20mV , C _L = 15pF		45	50	
			Input overdrive = 100mV , C _L = 15pF		42	50	
			T _A = −40°C to 125°C			55	
	Propagation delay skew	Input overdrive = 20mV , C _L = 15pF			2		ns
t _R	Rise time	10% to 90%			2.9		ns
t _F	Fall time	10% to 90%			3.7		ns

6.7 Electrical Characteristics: $V_{CC} = 2.7V$

at $T_A = 25^\circ C$ and $V_{CC} = 2.7V$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET VOLTAGE						
V _{IO-TLV3201}	Input offset voltage TLV3201	V _{CM} = V _{CC} / 2		1	3	mV
		T _A = −40°C to 125°C			4	
V _{IO-TLV3202}	Input offset voltage TLV3202	V _{CM} = V _{CC} / 2		1	5	mV
		T _A = −40°C to 125°C			6	
dV _{OS} /dT	Input offset voltage drift	T _A = −40°C to 125°C		1	10	μV/°C
PSRR	Power-supply rejection ratio	V _{CM} = V _{CC} / 2, V _{CC} = 2.5V to 5.5V	65	85		dB
	Input hysteresis			1.2		mV
INPUT BIAS CURRENT						
I _{IB}	Input bias current	V _{CM} = V _{CC} / 2		1	50	pA
		T _A = −40°C to 125°C			5	nA
I _{IO}	Input offset current	V _{CM} = V _{CC} / 2		1	50	pA
		T _A = −40°C to 125°C			2.5	nA
INPUT VOLTAGE RANGE						
V _{CM}	Common-mode voltage	T _A = −40°C to 125°C	(V _{EE}) − 0.2		(V _{CC}) + 0.2	V
CMRR	Common-mode rejection ratio	−0.2V < V _{CM} < 2.9V	56	68		dB
INPUT IMPEDANCE						
	Common mode			10 ¹³ 2		Ω pF
	Differential			10 ¹³ 4		Ω pF
OUTPUT						
V _{OL}	Voltage output swing from lower rail	I _{SINK} = 4mA		230	260	mV
		T _A = −40°C to 125°C			325	
V _{OH}	Voltage output swing from upper rail	I _{SOURCE} = 4mA		210	250	mV
		T _A = −40°C to 125°C			350	
I _{SC}	Short-circuit current (per comparator)	I _{SC} sinking	13	19		mA
		T _A = −40°C to 125°C		See Figure 6-14		
		I _{SC} sourcing	15	21		
		T _A = −40°C to 125°C		See Figure 6-14		
POWER SUPPLY						
V _{CC}	Specified voltage		2.7		5.5	V
I _Q	Quiescent current	T _A = 25°C		36	46	μA
		T _A = −40°C to 125°C			60	

6.8 Switching Characteristics: $V_{CC} = 2.7V$

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PD}	Propagation delay time	Low to high		47	50	ns
				42	50	
					55	
		High to low		40	50	
				38	50	
					55	
	Propagation delay skew	Input overdrive = 20mV, $C_L = 15pF$		2		ns
t_R	Rise time	10% to 90%		4.8		ns
t_F	Fall time	10% to 90%		5.2		ns

6.9 Typical Characteristics

at $T_A = 25^\circ\text{C}$, $V_{CC} = 5\text{V}$, and input overdrive (V_{OD}) = 20mV (unless otherwise noted)

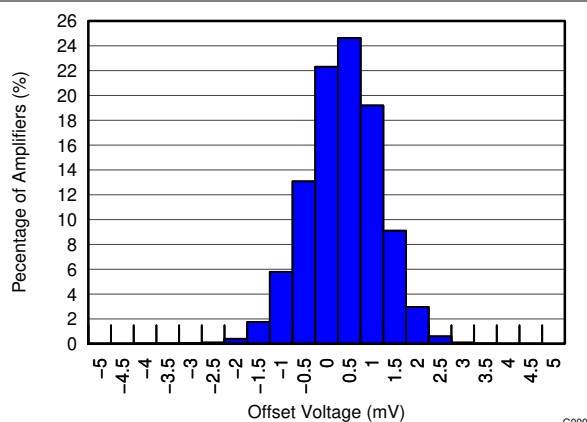


Figure 6-1. Offset Voltage Distribution

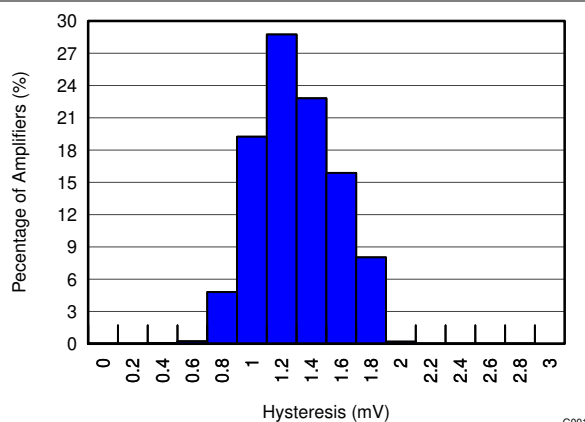


Figure 6-2. Hysteresis Distribution

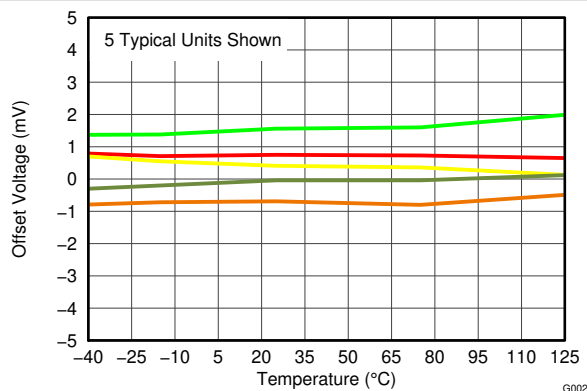


Figure 6-3. Offset Voltage vs Temperature

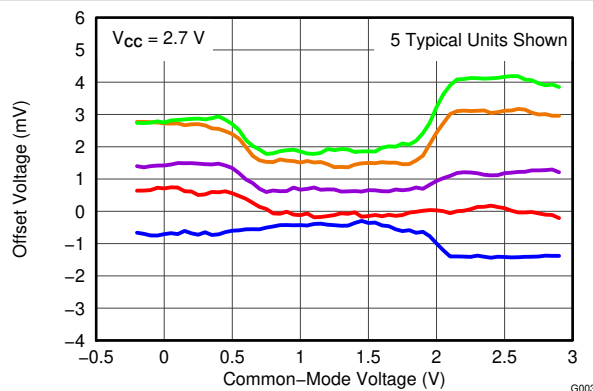


Figure 6-4. Offset Voltage vs Common-Mode Voltage

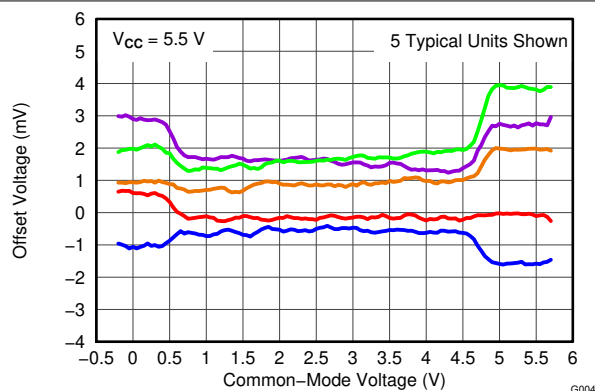


Figure 6-5. Offset Voltage vs Common-Mode Voltage

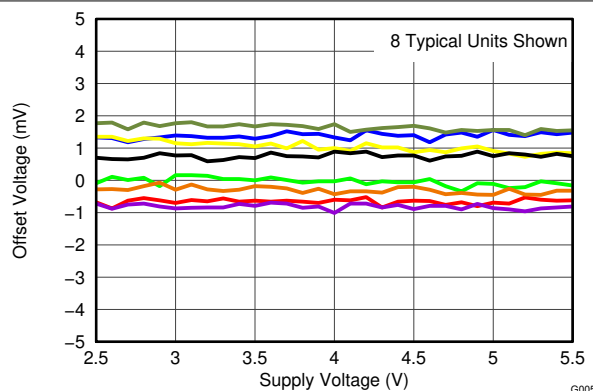


Figure 6-6. Offset Voltage vs Power Supply

6.9 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $V_{CC} = 5\text{V}$, and input overdrive (V_{OD}) = 20mV (unless otherwise noted)

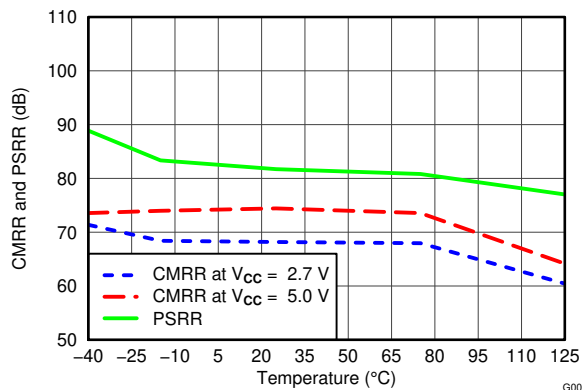


Figure 6-7. Common-Mode Rejection Ratio and Power-Supply Rejection Ratio vs Temperature

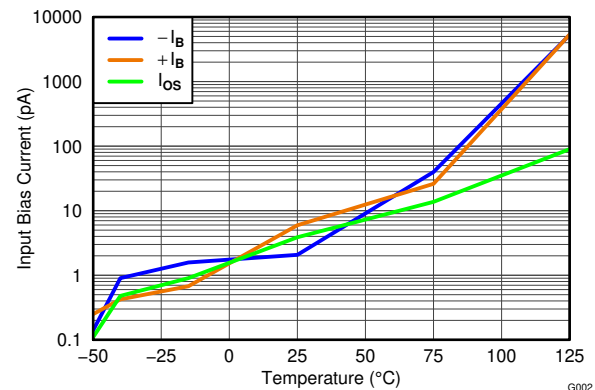


Figure 6-8. Input Bias Current and Input Offset Current vs Temperature

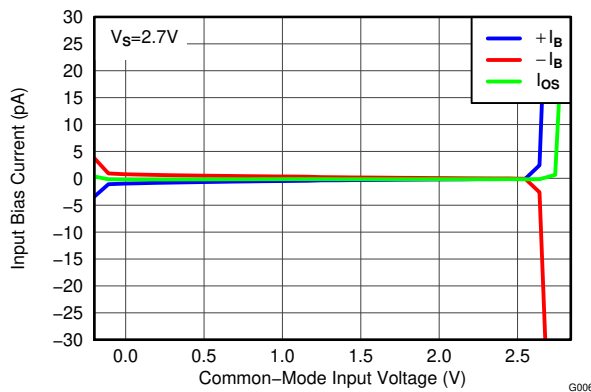


Figure 6-9. Input Bias Current and Input Offset Current vs Common-Mode Input Voltage

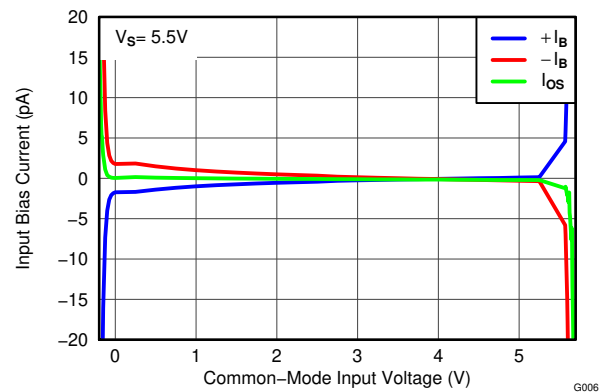


Figure 6-10. Input Bias Current and Input Offset Current vs Common-Mode Input Voltage

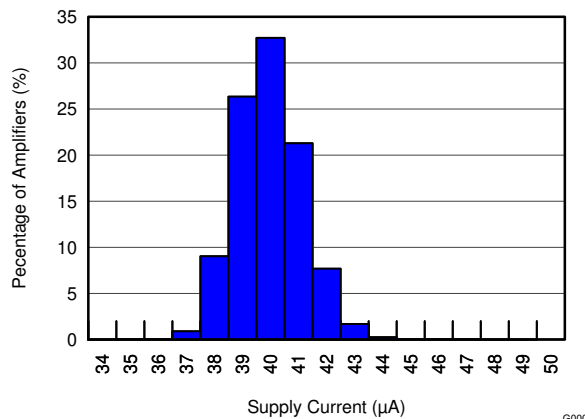


Figure 6-11. Quiescent Current Distribution

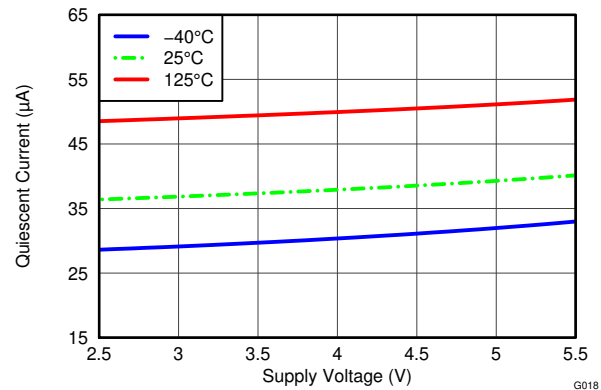


Figure 6-12. Quiescent Current vs Supply Voltage

6.9 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $V_{CC} = 5\text{V}$, and input overdrive (V_{OD}) = 20mV (unless otherwise noted)

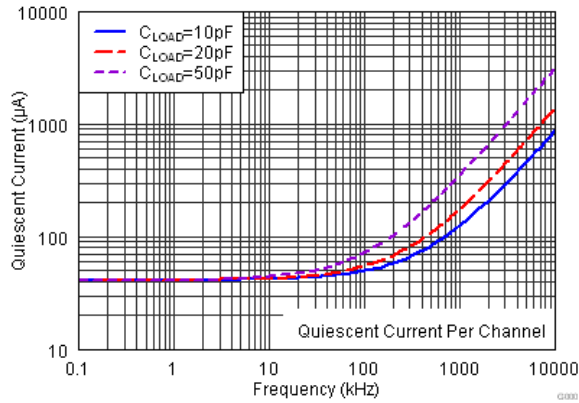


Figure 6-13. Quiescent Current vs Switching Frequency

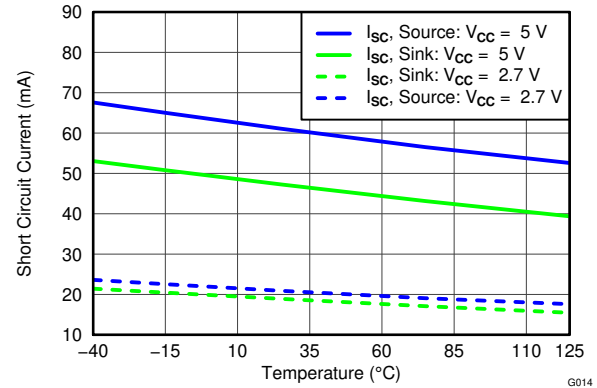


Figure 6-14. Short-Circuit Current vs Temperature

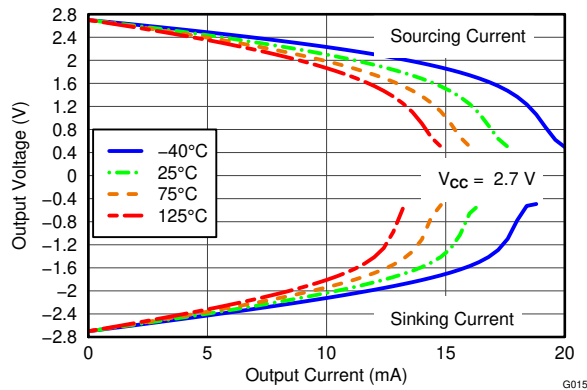


Figure 6-15. Output Voltage vs Output Current

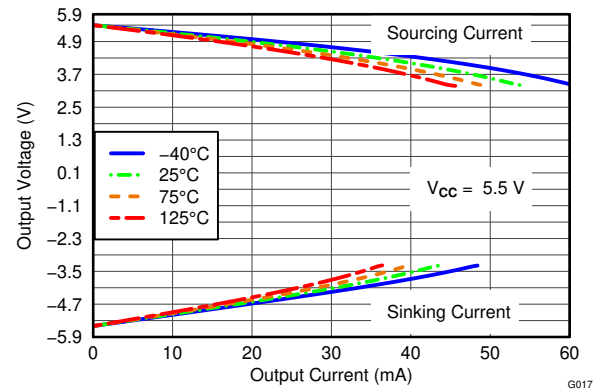


Figure 6-16. Output Voltage vs Output Current

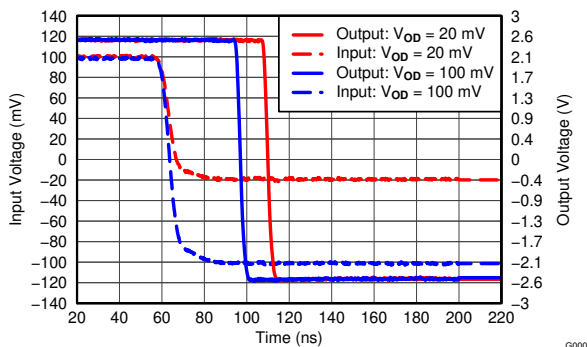


Figure 6-17. Propagation Delay Falling Edge

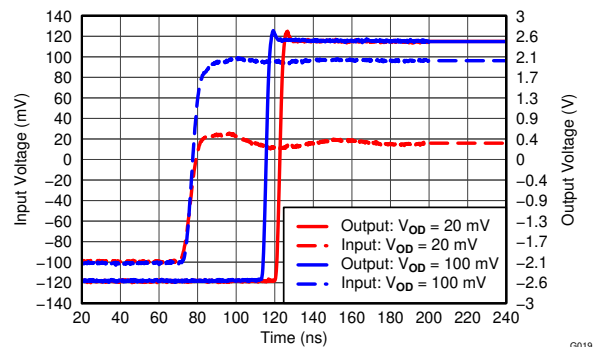


Figure 6-18. Propagation Delay Rising Edge

6.9 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $V_{CC} = 5\text{V}$, and input overdrive (V_{OD}) = 20mV (unless otherwise noted)

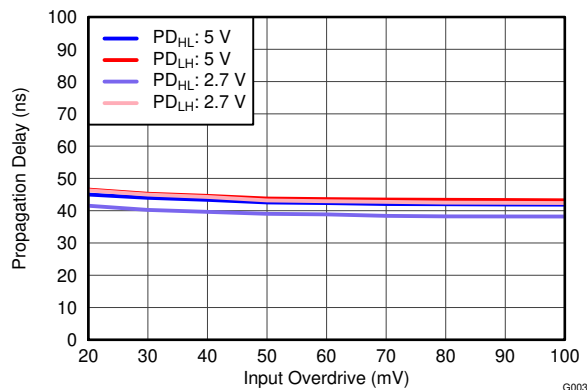


Figure 6-19. Propagation Delay vs Input Overdrive

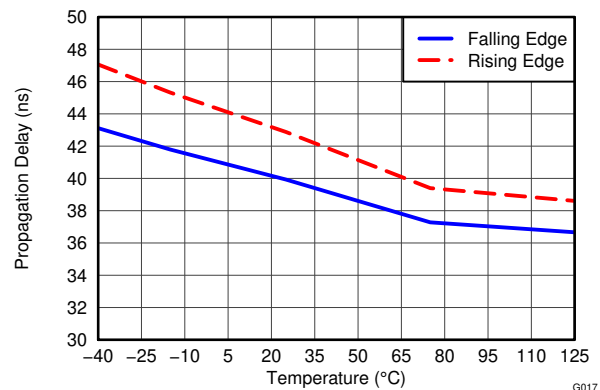


Figure 6-20. Propagation Delay vs Temperature

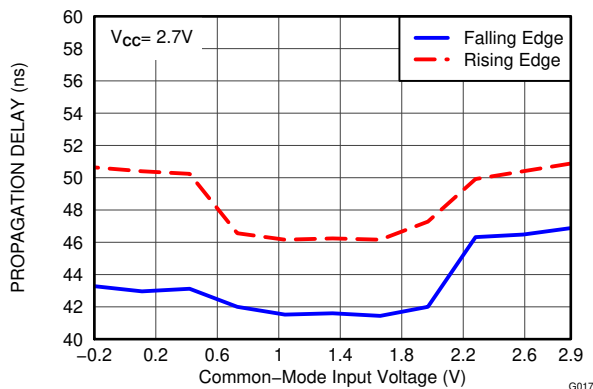


Figure 6-21. Propagation Delay vs Common-Mode Voltage

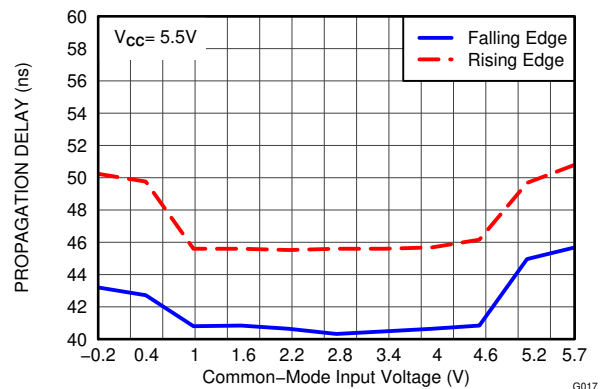


Figure 6-22. Propagation Delay vs Common-Mode Voltage

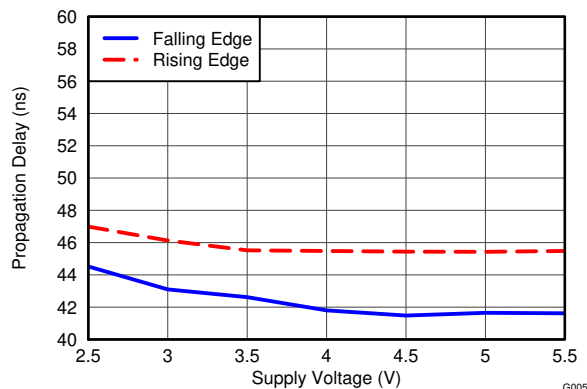


Figure 6-23. Propagation Delay vs Supply Voltage

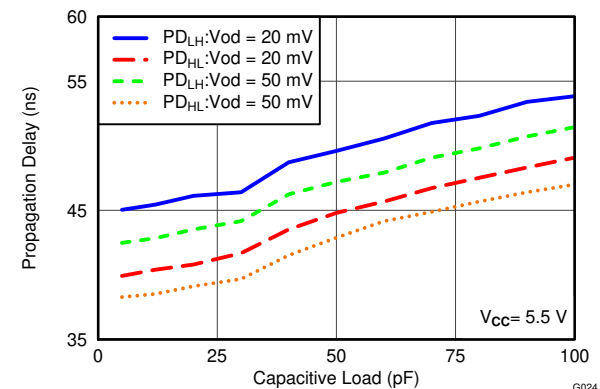


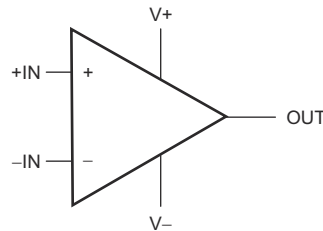
Figure 6-24. Propagation Delay vs Capacitive Load

7 Detailed Description

7.1 Overview

The TLV3201 and TLV3202 devices feature 40ns response time, and include 1.2mV of internal hysteresis for improved noise immunity with an input common-mode range that extends 0.2V beyond the power-supply rails.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 Operating Voltage

The TLV3201 and TLV3202 comparators are specified for use on a single supply from 2.7V to 5.5V (or a dual supply from $\pm 1.35\text{V}$ to $\pm 2.75\text{V}$) over a temperature range of -40°C to 125°C . The device continues to function below this range, but performance is not specified.

7.3.2 Input Overvoltage Protection

The device inputs are protected by electrostatic discharge (ESD) diodes that conduct if the input voltages exceed the power supplies by more than approximately 300mV. Momentary voltages greater than 300mV beyond the power supply can be tolerated if the input current is limited to 10mA. This limiting is easily accomplished with a small input resistor in series with the input to the comparator.

7.4 Device Functional Modes

The device is fully functional when powered by rail-to-rail supply voltage greater than 2.7V. The device is off at any voltages below 2.7V.

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

The TLV3201 and TLV3202 are single- and dual-supply (respectively), push-pull comparators featuring 40ns of propagation delay on only 40µA of supply current. This combination of fast response time and minimal power consumption make the TLV3201 and TLV3202 excellent comparators for portable, battery-powered applications as well as fast-switching threshold detection such as pulse-width modulation (PWM) output monitors and zero-cross detection.

8.1.1 Comparator Inputs

The TLV3201 and TLV3202 are rail-to-rail input comparators, with an input common-mode range that exceeds the supply rails by 200mV for both positive and negative supplies. The devices are specified from 2.7V to 5.5V, with room temperature operation from 2.5V to 5.5V. The TLV3201 and TLV3202 are designed to prevent phase inversion when the input pins exceed the supply voltage. Figure 8-1 shows the TLV320x response when input voltages exceed the supply, resulting in no phase inversion.

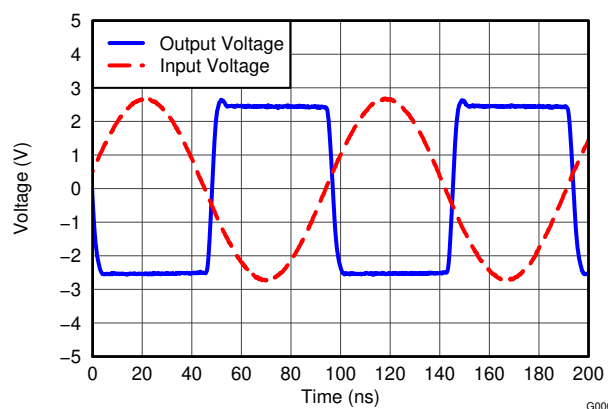
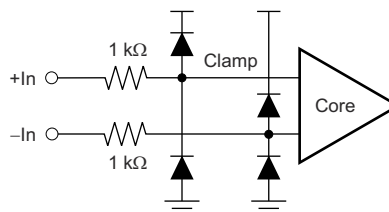


Figure 8-1. No Phase Inversion: Comparator Response to Input Voltage (Propagation Delay Included)

The electrostatic discharge (ESD) protection input structure of two back-to-back diodes and 1kΩ series resistors are used to limit the differential input voltage applied to the precision input of the comparator by clamping input voltages that exceed V_{CC} beyond the specified operating conditions. If potential overvoltage conditions that exceed absolute maximum ratings are present, the addition of external bypass diodes and resistors is recommended, as shown in Figure 8-2. Large differential voltages greater than the supply voltage must be avoided to prevent damage to the input stage.



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Figure 8-2. TLV3201 Equivalent Input structure

8.1.2 External Hysteresis

The TLV3201 and TLV3202 have a hysteresis transfer curve (shown in Figure 8-3) that is a function of three components: V_{TH} , V_{OS} , and V_{HYST} .

- V_{TH} : the actual set voltage or threshold trip voltage
- V_{OS} : the internal offset voltage between V_{IN+} and V_{IN-} . This voltage is added to V_{TH} to form the actual trip point at which the comparator must respond to change output states.
- V_{HYST} : internal hysteresis (or trip window) that is designed to reduce comparator sensitivity to noise.

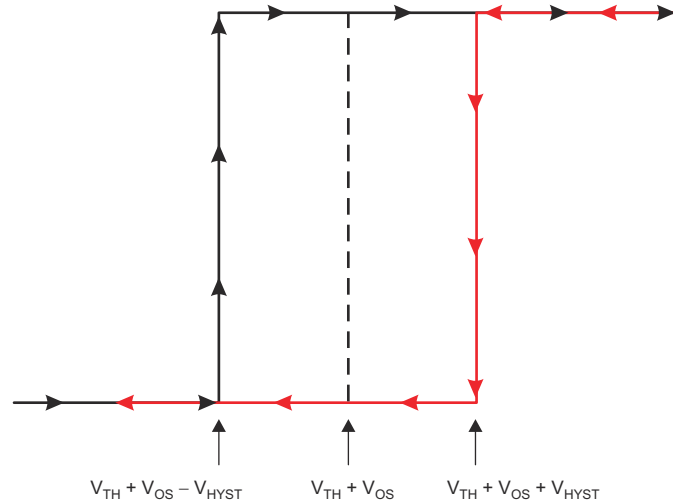


Figure 8-3. TLV320x Hysteresis Transfer Curve

8.1.2.1 Inverting Comparator with Hysteresis

The inverting comparator with hysteresis requires a three-resistor network that is referenced to the comparator supply voltage (V_{CC}), as shown in Figure 8-4. When V_{IN} at the inverting input is less than V_A , the output voltage is high (for simplicity, assume V_O switches as high as V_{CC}). The three network resistors can be represented as $R1 \parallel R3$ in series with $R2$. The lower input trip voltage (V_{A1}) is defined by Equation 1.

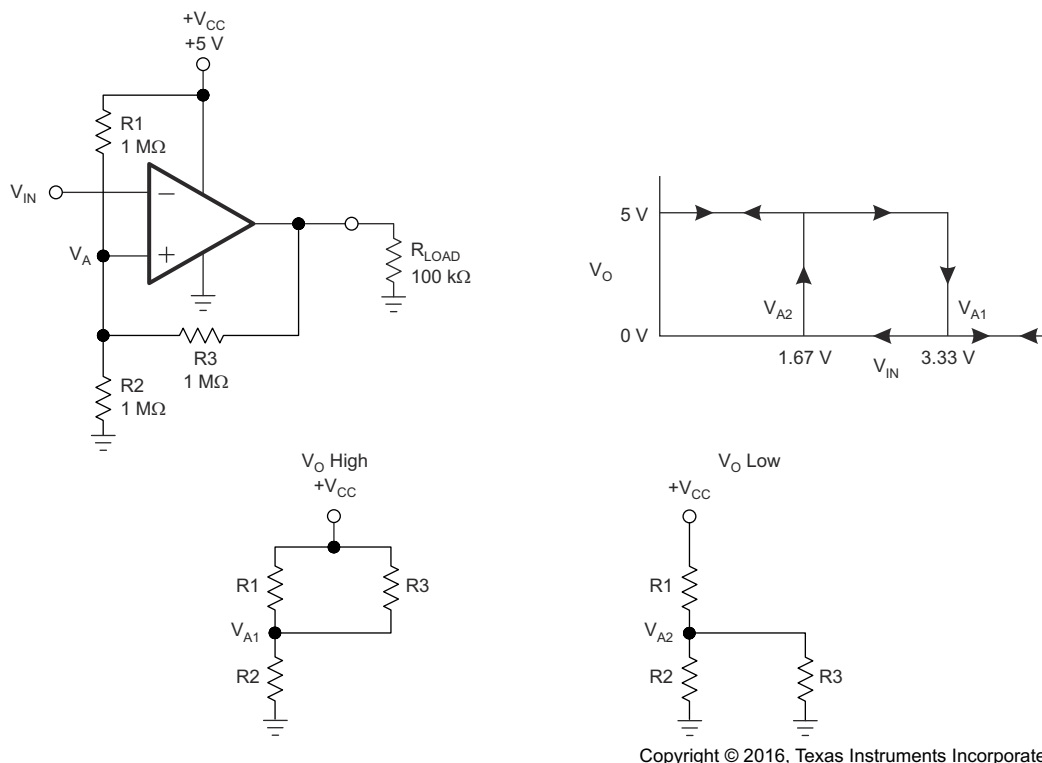
$$V_{A1} = V_{CC} \times \frac{R2}{(R1 \parallel R3) + R2} \quad (1)$$

When V_{IN} is greater than $[V_A \times (V_{IN} > V_A)]$, the output voltage is low, very close to ground. In this case, the three network resistors can be presented as $R2 \parallel R3$ in series with $R1$. The upper trip voltage (V_{A2}) is defined by Equation 2.

$$V_{A2} = V_{CC} \times \frac{R2 \parallel R3}{R1 + (R2 \parallel R3)} \quad (2)$$

The total hysteresis provided by the network is defined by Equation 3.

$$\Delta V_A = V_{A1} - V_{A2} \quad (3)$$



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Figure 8-4. TLV3201 in Inverting Configuration With Hysteresis

8.1.2.2 Noninverting Comparator With Hysteresis

A noninverting comparator with hysteresis requires a two-resistor network, as shown in [Figure 8-5](#), and a voltage reference (V_{REF}) at the inverting input. When V_{IN} is low, the output is also low. For the output to switch from low to high, V_{IN} must rise up to V_{IN1}. V_{IN1} is calculated by [Equation 4](#).

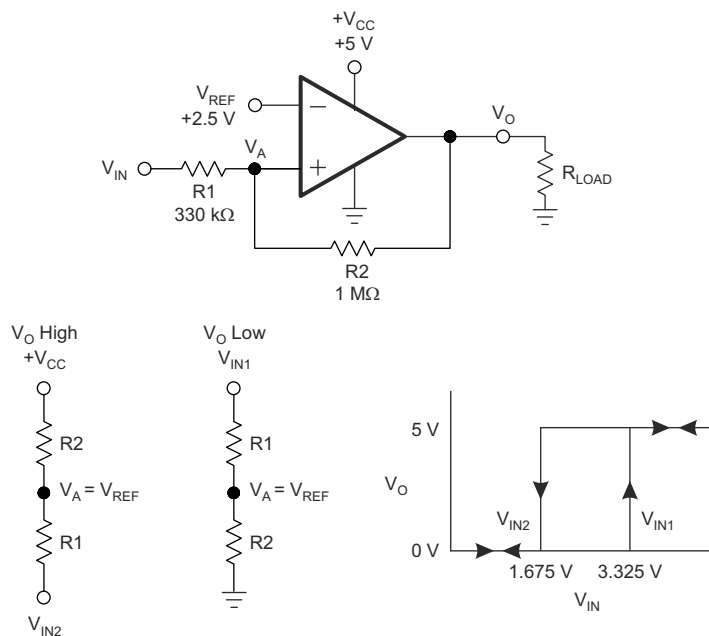
$$V_{IN1} = R1 \times \frac{V_{REF}}{R2} \times V_{REF} \quad (4)$$

When V_{IN} is high, the output is also high. In order for the comparator to switch back to a low state, V_{IN} must equal V_{REF} before V_A is again equal to V_{REF}. V_{IN} can be calculated by [Equation 5](#).

$$V_{IN2} = \frac{V_{REF} (R1 + R2) - V_{CC} \times R1}{R2} \quad (5)$$

The hysteresis of this circuit is the difference between V_{IN1} and V_{IN2}, as defined by [Equation 6](#).

$$\Delta V_{IN} = V_{CC} \times \frac{R1}{R2} \quad (6)$$



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Figure 8-5. TLV3201 in Noninverting Configuration With Hysteresis

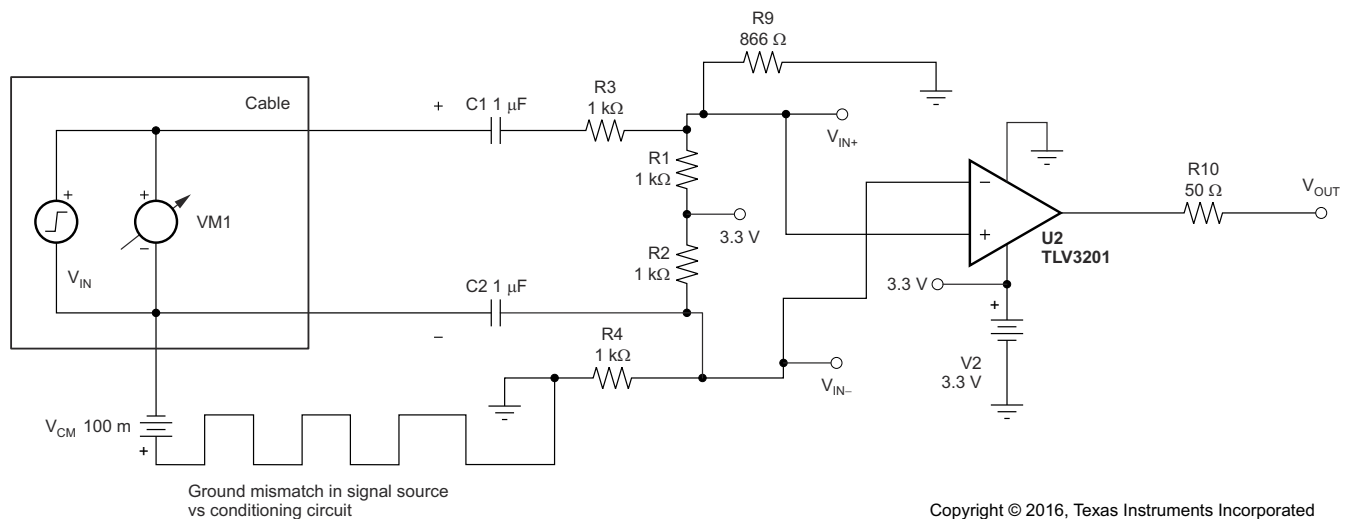
8.1.3 Capacitive Loads

The TLV3201 and TLV3202 feature a push-pull output. When the output switches, there is a direct path between V_{CC} and ground, causing increased output sinking or sourcing current during the transition. Following the transition the output current decreases and supply current returns to 40 μ A, thus maintaining low power consumption. Under reasonable capacitive loads, the TLV3201 and TLV3202 maintain specified propagation delay (see [Section 6.9](#)), but excessive capacitive loading under high switching frequencies may increase supply current, propagation delay, or induce decreased slew rate.

8.2 Typical Applications

8.2.1 TLV3201 Configured as an AC-Coupled Comparator

One of the benefits of AC coupling a single-supply comparator circuit is that it can block dc offsets induced by ground-loop offsets that could potentially produce either a false trip or a common-mode input violation. [Figure 8-6](#) shows the TLV3201 configured as an ac-coupled comparator.



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Figure 8-6. TLV3201 Configured as an AC-Coupled Comparator (Schematic)

8.2.1.1 Design Requirements

Design requirements include:

- Ability to tolerate up to $\pm 100\text{mV}$ of common-mode signal.
- Trigger only on AC signals (such as zero-cross detection).

8.2.1.2 Detailed Design Procedure

Design analysis:

- AC-coupled, high-pass frequency
 - Large capacitors require longer start-up time from device power on
 - Use $1\mu\text{F}$ capacitor to achieve high-pass frequency of approximately 159Hz
 - For high-pass equivalent, use $C_{\text{IN}} = 0.5\mu\text{F}$, $R_{\text{IN}} = 2\text{k}\Omega$
1. Set up input dividers initially for one-half supply (to be in center of acceptable common-mode range).
 2. Adjust either divider slightly upwards or downwards as desired to establish quiescent output condition.
 3. Select coupling capacitors based on lowest expected frequency.

8.2.1.3 Application Curve

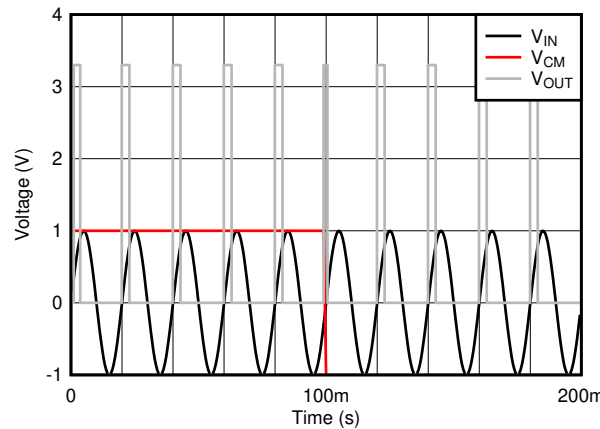
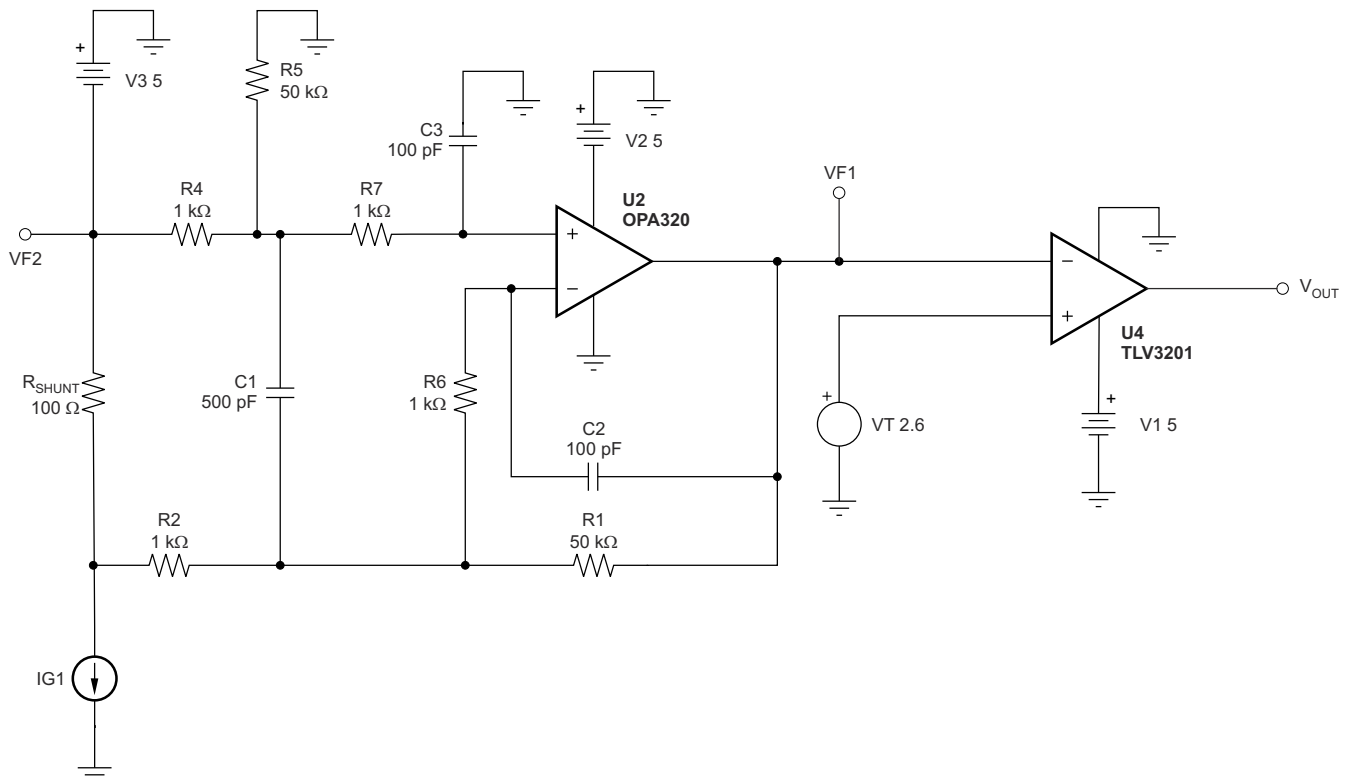


Figure 8-7. AC-Coupled Comparator Results

8.2.2 TLV3201 and OPA320 Configured as a Fast-Response Output Current Monitor

Figure 8-8 shows a single-supply current monitor configured as a difference amplifier with a gain of 50. The OPA320 was chosen for this circuit because of its gain bandwidth (20 MHz), which allows higher speed triggering and monitoring of the current across the shunt resistor followed by the fast response of the TLV3201.

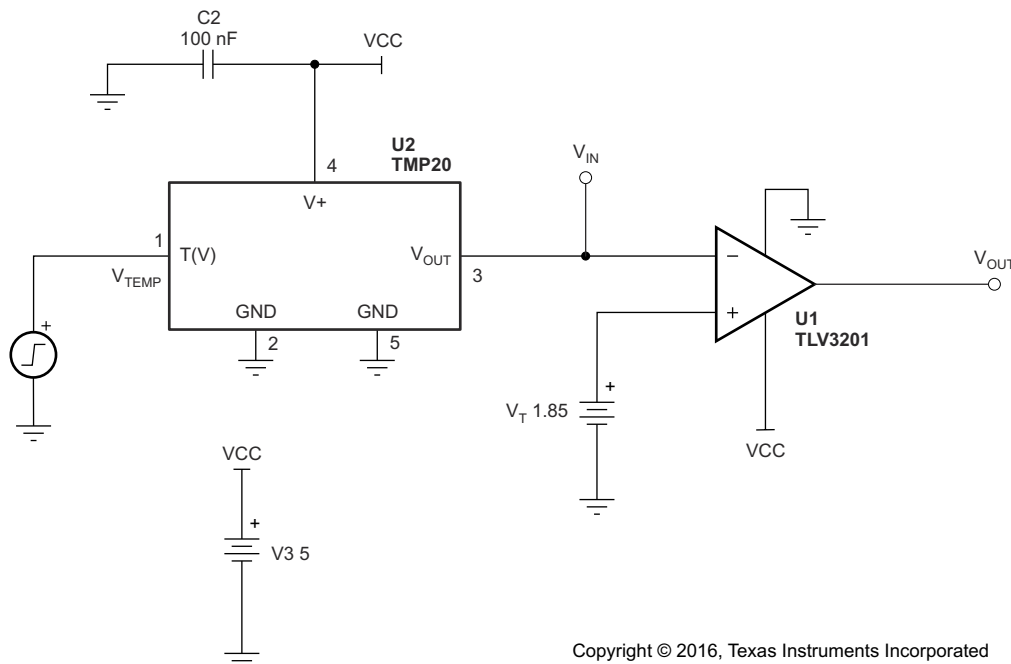


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Figure 8-8. TLV3201 and OPA320 Configured as a Fast-Response Output Current Monitor

8.2.3 TLV3201 and TMP20 Configured as a Precision Analog Temperature Switch

Figure 8-9 shows the TMP20 and TLV3201 designed as a high-speed temperature switch. The TMP20 is an analog output temperature sensor where output voltage decreases with temperature. The comparator output is tripped when the output reaches a critical trip threshold.



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Figure 8-9. TLV3201 and TMP20 Configured as a Precision Analog Temperature Switch

8.3 Power Supply Recommendations

The TLV3201 and TLV3202 comparators are specified for use on a single supply from 2.7V to 5.5V (or a dual supply from $\pm 1.35\text{V}$ to $\pm 2.75\text{V}$) over a temperature range of -40°C to 125°C . The device continues to function below this range, but performance is not specified. Place bypass capacitors close to the power-supply pins to reduce noise coupling in from noisy or high-impedance power supplies. For more detailed information on bypass capacitor placement, see [Section 8.4.1](#).

8.4 Layout

8.4.1 Layout Guidelines

The TLV3201 and TLV3202 are fast-switching, high-speed comparators and require high-speed layout considerations. For best results, maintain the following layout guidelines.

- Use a printed-circuit board (PCB) with a good, unbroken low-inductance ground plane.
- Place a decoupling capacitor (0.1 μF ceramic, surface-mount capacitor) as close as possible to V_{CC} .
- On the inputs and the output, keep lead lengths as short as possible to avoid unwanted parasitic feedback around the comparator. Keep inputs away from the output.
- Solder the device directly to the PCB rather than using a socket.
- For slow-moving input signals, take care to prevent parasitic feedback. A small capacitor (1000pF or less) placed between the inputs can help eliminate oscillations in the transition region. This capacitor causes some degradation to propagation delay when the impedance is low. The topside ground plane runs between the output and inputs.
- The ground pin ground trace runs under the device up to the bypass capacitor, shielding the inputs from the outputs.

8.4.2 Layout Example

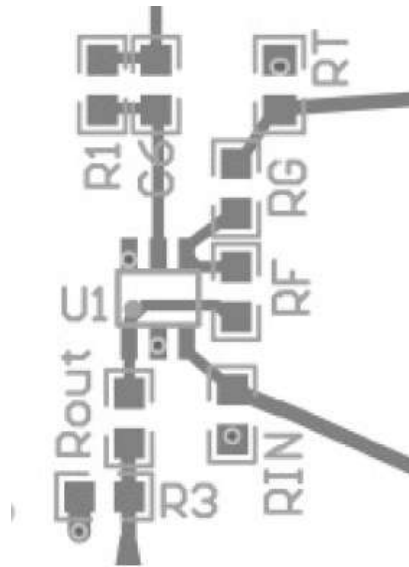


Figure 8-10. TLV320x Board Layout Example

9 Device and Documentation Support

9.1 Device Support

9.1.1 Development Support

9.1.1.1 TINA-TI™ (Free Software Download)

TINA-TI™ is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional dc, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a [free download](#) from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic quick-start tool.

Note

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the [TINA-TI folder](#).

9.1.1.2 Universal Op Amp EVM

The [Universal Op Amp EVM](#) is a series of general-purpose, blank circuit boards that simplify prototyping circuits for a variety of IC package types. The evaluation module board design allows many different circuits to be constructed easily and quickly. Five models are offered, with each model intended for a specific package type. PDIP, SOIC, MSOP, TSSOP and SOT23 packages are all supported.

Note

These boards are unpopulated, so users must provide their own ICs. TI recommends requesting several op amp device samples when ordering the Universal Op Amp EVM.

9.1.1.3 TI Precision Designs

TI Precision Designs are analog solutions created by TI's precision analog applications experts and offer the theory of operation, component selection, simulation, complete PCB schematic and layout, bill of materials, and measured performance of many useful circuits. TI Precision Designs are available online at <http://www.ti.com/ww/en/analog/precision-designs/>.

9.1.1.4 WEBENCH® Filter Designer

[WEBENCH® Filter Designer](#) is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

Available as a web-based tool from the WEBENCH® Design Center, [WEBENCH® Filter Designer](#) allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

9.2 Documentation Support

9.2.1 Related Documentation

The following documents are relevant to using the TLV320x, and recommended for reference. All are available for download at www.ti.com unless otherwise noted.

- [Frequency Dithering With the UCC28950 and TLV3201](#) (SLUA646)
- [Frequency Dithering with the UCC28180 and TLV3201](#) (SLUA704)
- [Comparator with Hysteresis Reference Design](#) (TIDU020)

9.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

9.4 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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9.5 Trademarks

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9.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

9.7 Glossary

[TI Glossary](#)

This glossary lists and explains terms, acronyms, and definitions.

10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision B (December 2016) to Revision C (June 2022)	Page
• Corrected transposed TLV3202 D and DGK columns in Thermal Metric table.....	4
• Corrected typo in DGK Tja from 146.3 to 143.6 for DGK in Thermal Metric table.....	4
• Added Input Offset Voltage for TLV3201 for 5V.....	5
• Added 5V Switching Specifications table.....	5
• Added Input Offset Voltage for TLV3201 for 2.7V.....	6

Changes from Revision A (June 2012) to Revision B (December 2016)	Page
• Added <i>ESD Ratings</i> table, <i>Feature Description</i> section, <i>Device Functional Modes</i> , <i>Application and Implementation</i> section, <i>Power Supply Recommendations</i> section, <i>Layout</i> section, <i>Device and Documentation Support</i> section, and <i>Mechanical, Packaging, and Orderable Information</i> section	1
• Deleted <i>Ordering Information</i> table; see <i>Package Option Addendum</i> at the end of the data sheet.....	1

Changes from Revision * (March 2012) to Revision A (June 2012)	Page
• Changed product status from Production Data to Mixed Status.....	1
• Added dual channel device.....	1

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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